



13MHz, Rail-to-Rail I/O CMOS Operational Amplifier

1 FEATURES

- Qualified for Automotive Applications
- AEC-Q100 Qualified with the Grade 1
- HIGH GAIN BANDWIDTH: 13MHz
- RAIL-TO-RAIL INPUT AND OUTPUT ±0.5mV Typical Vos
- INPUT VOLTAGE RANGE: -0.1V to +5.6V with Vs = 5.5V
- SUPPLY RANGE: +2.7V to +5.5V
- SPECIFIED UP TO +125°C
- Micro SIZE PACKAGES: SOP14, TSSOP14, QFN3X3-16

2 APPLICATIONS

- SENSORS
- PHOTODIODE AMPLIFICATION
- ACTIVE FILTERS
- TEST EQUIPMENT
- DRIVING A/D CONVERTERS

3 DESCRIPTIONS

The RS724-Q1 families of products offer low voltage operation and rail-to-rail input and output, as well as excellent speed/power consumption ratio, providing an excellent bandwidth (13MHz) and slew rate of 8V/us. The op-amps are unity gain stable and feature an ultra-low input bias current.

The devices are ideal for sensor interfaces, active filters and portable applications. The RS724-Q1 families of operational amplifiers are specified at the full temperature range of -40°C to +125°C under single or dual power supplies of 2.7V to 5.5V.

Device Information(1)

PART NUMBER	PACKAGE	BODY SIZE (NOM)
	SOP14	8.65mm×3.90mm
RS724-Q1	TSSOP14	5.00mm×4.40mm
	QFN3X3-16	3.00mm×3.00mm

(1) For all available packages, see the orderable addendum at the end of the data sheet.



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4 Revision History

Note: Page numbers for previous revisions may different from page numbers in the current version.

	VERSION	Change Date	Change Item
	A.1	2023/08/16	Initial version completed
ſ	A.1.1	2024/03/07	Modify packaging naming



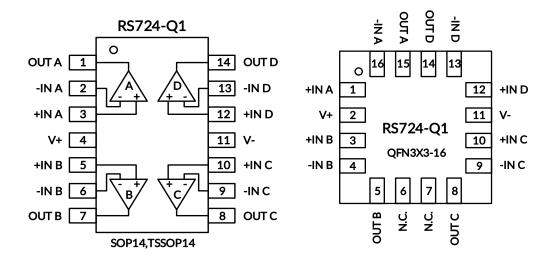
5 PACKAGE/ORDERING INFORMATION (1)

Orderable Device	Package Type	Pin	Channel	Lead finish/Ball material ⁽²⁾	MSL Peak Temp ⁽³⁾	Op Temp(°C)	Device Marking	Package Qty
RS724XP -Q1	SOP14	14	4	NIPDAUAG	MSL1-260°- Unlimited	-40°C ~125°C	RS724	Tape and Reel,4000
RS724XQ -Q1	TSSOP14	14	4	NIPDAUAG	MSL1-260°- Unlimited	-40°C ~125°C	RS724	Tape and Reel,4000
RS724XT QC16-Q1	QFN3X3- 16	16	4	NIPDAUAG	MSL1-260°- Unlimited	-40°C ~125°C	RS724	Tape and Reel,5000

- (1) This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the right-hand navigation.
- (2) Lead finish/Ball material. Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.
- (3) MSL Peak Temp. The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.
- (4) There may be additional marking, which relates to the lot trace code information (data code and vendor code), the logo or the environmental category on the device.



6 Pin Configuration and Functions (Top View)



Pin Description

NANAT	PIN		I/O (1)	DESCRIPTION
NAME	SOP14/TSSOP14	QFN3X3-16	1/01-/	DESCRIPTION
-INA	2	16	I	Inverting input, channel A
+INA	3	1	I	Noninverting input, channel A
-INB	6	4	I	Inverting input, channel B
+INB	5	3	I	Noninverting input, channel B
-INC	9	9	I	Inverting input, channel C
+INC	10	10	I	Noninverting input, channel C
-IND	13	13	I	Inverting input, channel D
+IND	12	12	I	Noninverting input, channel D
OUTA	1	15	0	Output, channel A
OUTB	7	5	0	Output, channel B
OUTC	8	8	0	Output, channel C
OUTD	14	14	0	Output, channel D
V-	11	11	-	Negative (lowest) power supply
V+	4	2	-	Positive (highest) power supply

⁽¹⁾ I = Input, O = Output.



7 SPECIFICATIONS

7.1 Absolute Maximum Ratings

Over operating free-air temperature range (unless otherwise noted) (1)

			MIN	MAX	UNIT	
	Supply, V _S =(V+) - (V-)			7		
Voltage	Signal input pin ⁽²⁾		(V-)-0.5	(V+) +0.5	V	
	Signal output pin ⁽³⁾		(V-)-0.5	(V+) +0.5		
	Signal input pin ⁽²⁾		-10	10	mA	
Current	Signal output pin ⁽³⁾		-200	200	mA	
	Output short-circuits (4)		Cont	Continuous		
	Package thermal impedance (5)	SOP14		105		
θJA		TSSOP14		90	°C/W	
		QFN3X3-16		70		
	Operating range, T _A	-40	125			
Temperature	Junction, T _J ⁽⁶⁾		-40	150	°C	
	Storage, T _{stg}	-65	150			

⁽¹⁾ Stresses above these ratings may cause permanent damage. Exposure to absolute maximum conditions for extended periods may degrade device reliability. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those specified is not implied.

- (4) Short-circuit to ground, one amplifier per package.
- (5) The package thermal impedance is calculated in accordance with JESD-51.
- (6) The maximum power dissipation is a function of $T_{J(MAX)}$, $R_{\theta JA}$, and T_A . The maximum allowable power dissipation at any ambient temperature is $P_D = (T_{J(MAX)} T_A) / R_{\theta JA}$. All numbers apply for packages soldered directly onto a PCB.

7.2 ESD Ratings

The following ESD information is provided for handling of ESD-sensitive devices in an ESD protected area only.

			VALUE	UNIT
		Human-Body Model (HBM), per AEC Q100-002 (1)	±2000	V
V _(ESD)	Electrostatic discharge	Charged-Device Model (CDM), per AEC Q100-011	±500	
		Latch-Up (LU), per AEC Q100-004	±100	mA

⁽¹⁾ AEC Q100-002 indicates that HBM stressing shall be in accordance with the ANSI/ESDA/JEDEC JS-001 specification.



ESD SENSITIVITY CAUTION

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

7.3 Recommended Operating Conditions

Over operating free-air temperature range (unless otherwise noted).

		MIN	NOM	MAX	UNIT
Supply voltage, V _S = (V+) - (V-)	Single-supply	2.7		5.5	V
Supply Voltage, Vs= (V+) - (V-)	Dual-supply	±1.35		±2.75	V

⁽²⁾ Input terminals are diode-clamped to the power-supply rails. Input signals that can swing more than 0.5V beyond the supply rails should be current-limited to 10mA or less.

⁽³⁾ Output terminals are diode-clamped to the power-supply rails. Output signals that can swing more than 0.5V beyond the supply rails should be current-limited to ±200mA or less.



7.4 ELECTRICAL CHARACTERISTICS

(At T_A = +25°C, V_S =2.7V to 5.5V, R_L = 10k Ω connected to $V_S/2$, and V_{OUT} = $V_S/2$, Full $^{(9)}$ = -40°C to +125°C, unless otherwise noted.) $^{(1)}$

	DADAMETED	COMPITIONS	_	RS724-Q1			
	PARAMETER	CONDITIONS	T,	MIN (2)	TYP (3)	MAX (2)	UNIT
POWER	RSUPPLY						
Vs	Operating Voltage Range		25°C	2.7		5.5	V
10	0 : 10 1/4 1/6)/ .0.5)/ l .0. 4	25°C		1.15	1.35	
IQ	Quiescent Current/Amplifier	V _S =±2.5V, Io=0mA	Full			1.85	mA
DCDD	D C D : 1: D :	\\ 0.7\\\ 5.5\\	25°C	75	93		
PSRR	Power-Supply Rejection Ratio	V _S =2.7V to 5.5V	Full	70			dB
INPUT							
Vos	Input Offset Voltage	V _{CM} =V _S /2	25°C	-1.5	±0.5	1.5 m\	
VOS	Input Offset Voltage	VCM-VS/2	Full	-3		3	IIIV
Vos Tc	Input Offset Voltage Average Drift		Full		±2.6		uV/°0
ın	L	V V /2	25°C		±1	±10	рA
IB	Input Bias Current (4) (5)	V _{CM} =V _S /2	Full		0.5	±10	nA
	L	V V /2	25°C		±1	±10	рA
los	Input Offset Current (5)	V _{CM} =V _S /2	Full			±10	nA
V _{СМ}	Common-Mode Voltage Range	V _S = 5.5V	25°C	-0.1		5.6	V
CMDD	Common-Mode Rejection Ratio	V _S = 5.5V, V _{CM} =-0.1V to 3.5V V _S = 5.5V, V _{CM} =-0.1V to 5.6V	25°C	73	90		dB
			Full	70			
CMRR			25°C	60	77		
			Full	59			
OUTPU	т						
۸	Open Lean Voltage Cain	R _L =10KΩ, Vo=(V-)	25°C	110	127		٩b
Aol	Open-Loop Voltage Gain	+0.1V to (V+)-0.1V	Full	94			dB
	Output Swing from Rail	V _S =±2.5V, R _L =10KΩ	25°C		10	20	mV
	Output Swing from Kaii	V5-12.5V, KL-10K22	Full			25	IIIV
Іоит	Output Short-Circuit Current ⁽⁶⁾	V _S =±2.5V, Vo=0V	25°C	±80	±150		mΛ
1001	(7)	VS-12.5V, VO-0V	Full	±60			mA
C_{LOAD}	Capacitive Load Drive		25°C		100		pF
FREQU	ENCY RESPONSE						
SR	Slew Rate (8)	G=+1, C _L =100pF	25°C		8		V/us
GBP	Gain-Bandwidth Product		25°C		13		MHz
ts	Settling Time,0.1%	V _S =±2.5V, G=+1, C _L =100pF, Step=2V	25°C		0.8		us
t_{OR}	Overload Recovery Time	V _{IN} ·Gain≥V _S , G=-10	25°C		0.4		us
NOISE							
En	Input Voltage Noise	f = 0.1Hz to 10Hz, V _S =±2.5V	25°C		5		uV _{PF}



- (1) Electrical table values apply only for factory testing conditions at the temperature indicated. Factory testing conditions result in very limited self-heating of the device.
- (2) Limits are 100% production tested at 25°C. Limits over the operating temperature range are ensured through correlations using statistical quality control (SQC) method.
- (3) Typical values represent the most likely parametric norm as determined at the time of characterization. Actual typical values may vary over time and will also depend on the application and configuration.
- (4) Positive current corresponds to current flowing into the device.
- (5) This parameter is ensured by design and/or characterization and is not tested in production.
- (6) The maximum power dissipation is a function of $T_{J(MAX)}$, $R_{\theta JA}$, and T_A . The maximum allowable power dissipation at any ambient temperature is PD = $(T_{J(MAX)} T_A) / R_{\theta JA}$. All numbers apply for packages soldered directly onto a PCB.
- (7) Short circuit test is a momentary test.
- (8) Number specified is the slower of positive and negative slew rates.
- (9) Specified by characterization only.



7.5 TYPICAL CHARACTERISTICS

NOTE: The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only.

At $T_A = +25^{\circ}C$, $V_S=5V$, $R_L = 10k\Omega$ connected to $V_S/2$, $V_{OUT} = V_S/2$, unless otherwise noted.

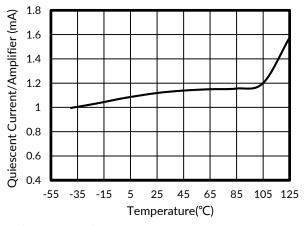
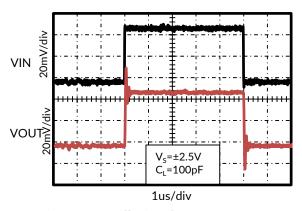


Figure 1. Quiescent Current vs Temperature

Figure 2. Input Bias Current vs Temperature



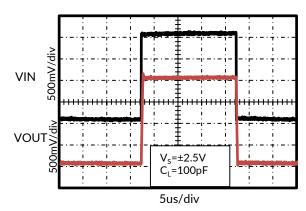
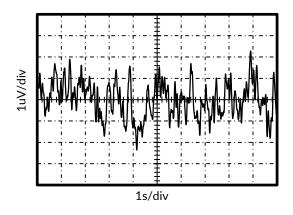


Figure 3. Small-Signal Step Response

Figure 4. Large-Signal Step Response



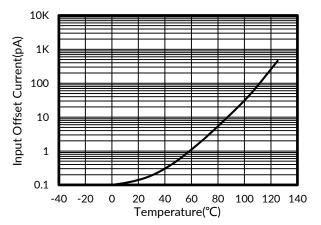


Figure 5. 0.1Hz to 10Hz Input Voltage Noise

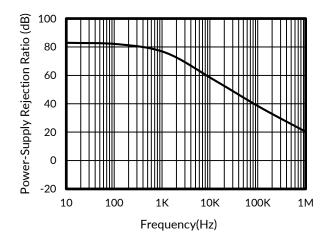
Figure 6. Input Offset Current vs Temperature



TYPICAL CHARACTERISTICS

NOTE: The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only.

At T_A = +25°C, V_S =5V, R_L = 10k Ω connected to $V_S/2$, V_{OUT} = $V_S/2$, unless otherwise noted.



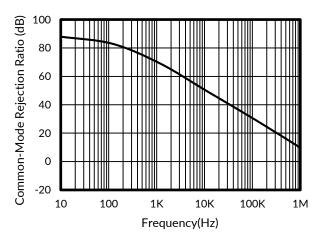


Figure 7. PSRR vs Frequency

Figure 8. CMRR vs Frequency



8 Detailed Description

8.1 Overview

The RS724-Q1 are high precision, rail-to-rail operational amplifiers that can be run from a single-supply voltage 2.7V to 5.5V ($\pm 1.35V$ to $\pm 2.75V$). Supply voltages higher than 7V (absolute maximum) can permanently damage the amplifier. Rail-to-rail input and output swing significantly increases dynamic range, especially in low-supply applications. Good layout practice mandates use of a 0.1uF capacitor place closely across the supply pins.

8.2 Phase Reversal Protection

The RS724-Q1 family has internal phase-reversal protection. Many op amps exhibit phase reversal when the input is driven beyond the linear common-mode range. This condition is most often encountered in noninverting circuits when the input is driven beyond the specified common-mode voltage range, causing the output to reverse into the opposite rail. The input of the RS724-Q1 prevents phase reversal with excessive common-mode voltage. Instead, the appropriate rail limits the output voltage. This performance is shown in figure 9.

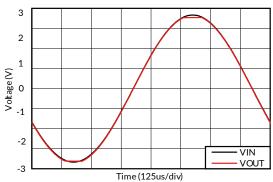


Figure 9. Output Waveform Devoid of Phase Reversal During an Input Overdrive Condition

8.3 EMIRR IN+ Test Configuration

Figure 10 shows the circuit configuration for testing the EMIRR IN+. An RF source is connected to the operational amplifier noninverting input pin using a transmission line. The operational amplifier is configured in a unity-gain buffer topology with the output connected to a low-pass filter (LPF) and a digital multimeter (DMM). A large impedance mismatch at the operational amplifier input causes a voltage reflection; however, this effect is characterized and accounted for when determining the EMIRR IN+. The resulting dc offset voltage is sampled and measured by the multimeter. The LPF isolates the multimeter from residual RF signals that can interfere with multimeter accuracy.

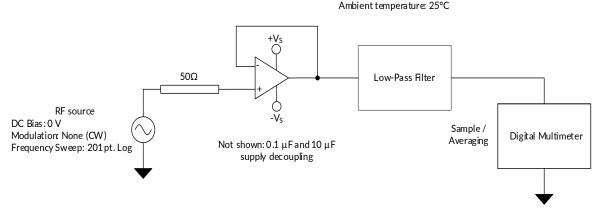


Figure 10. EMIRR IN+ Test Configuration Schematic

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9 Application and Implementation

Information in the following applications sections is not part of the RUNIC component specification, and RUNIC does not warrant its accuracy or completeness. RUNIC's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

9.1 APPLICATION NOTE

The RS724-Q1 are high precision, rail-to-rail operational amplifiers that can be run from a single-supply voltage 2.7V to 5.5V ($\pm 1.35V$ to $\pm 2.75V$). Supply voltages higher than 7V (absolute maximum) can permanently damage the amplifier. Rail-to-rail input and output swing significantly increases dynamic range, especially in low-supply applications. Good layout practice mandates use of a 0.1uF capacitor place closely across the supply pins.

Typical Applications 9.2 25-kHz Low-pass Filter

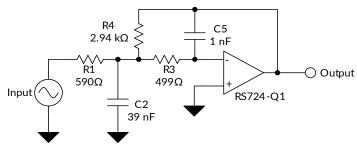


Figure 11. 25-kHz Low-Pass Filter

9.3 Design Requirements

Low-pass filters are commonly employed in signal processing applications to reduce noise and prevent aliasing. The RS724-Q1 devices are ideally suited to construct high-speed, high-precision active filters. Figure 11 shows a second-order, low-pass filter commonly encountered in signal processing applications.

Use the following parameters for this design example:

- Gain = 5 V/V (inverting gain)
- Low-pass cutoff frequency = 25 kHz
- Second-order Chebyshev filter response with 3-dB gain peaking in the passband.

9.4 Detailed Design Procedure

The infinite-gain multiple-feedback circuit for a low-pass network function is shown in Figure 11. Use Equation 1 to calculate the voltage transfer function.

$$\frac{\text{Output}}{\text{Input}}(s) = \frac{-1/R_1 R_3 C_2 C_5}{S^2 + (S/C_2)(1/R_1 + 1/R_3 + 1/R_4) + 1/R_3 R_4 C_2 C_5}$$
(1)

This circuit produces a signal inversion. For this circuit, the gain at dc and the low-pass cutoff frequency are calculated by Equation 2:

Gain =
$$\frac{R_4}{R_1}$$

 $f_c = \frac{1}{2\pi} \sqrt{(1/R_3 R_4 C_2 C_5)}$ (2)



9.5 Application Curve

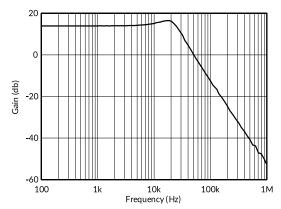


Figure 12. Low-pass filter transfer function



10 LAYOUTS

10.1 Layout Guidelines

Attention to good layout practices is always recommended. Keep traces short. When possible, use a PCB ground plane with surface-mount components placed as close to the device pins as possible. Place a 0.1uF capacitor closely across the supply pins.

These guidelines should be applied throughout the analog circuit to improve performance and provide benefits such as reducing the EMI susceptibility.

10.2 Layout Example

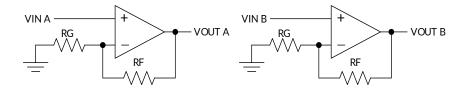


Figure 13. Schematic Representation

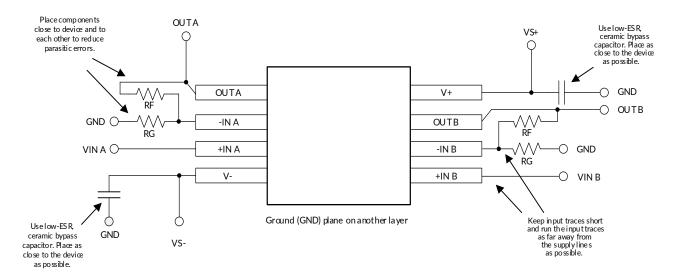
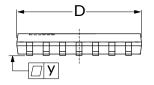


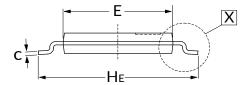
Figure 14. Layout Example

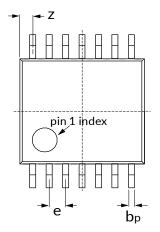
NOTE: Layout Recommendations have been shown for dual op-amp only, follow similar precautions for Single and four.

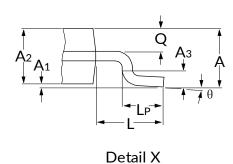


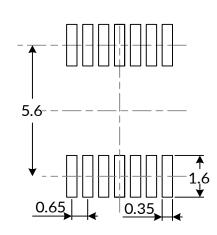
11 PACKAGE OUTLINE DIMENSIONS TSSOP14 (2)











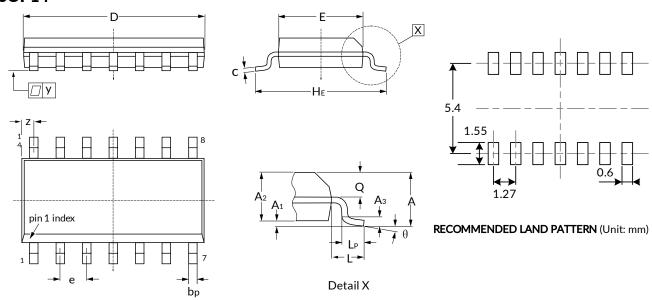
RECOMMENDED LAND PATTERN (Unit: mm)

C	Dimensions l	n Millimeters	Dimensions In Inches		
Symbol	Min	Max	Min	Max	
A ⁽¹⁾		1.100		0.043	
A ₁	0.050	0.150	0.002	0.006	
A ₂	0.800	0.950	0.031	0.037	
A ₃	0.	25	0.0)10	
bp	0.190	0.300	0.007	0.012	
С	0.100	0.200	0.004	0.008	
D ⁽¹⁾	4.900	5.100	0.193	0.201	
E ⁽¹⁾	4.300	4.500	0.169	0.177	
HE	6.200	6.600	0.244	0.260	
е	0.650		0.0)26	
L		1)39	
Lp	0.500	0.750	0.020	0.030	
Q	0.300	0.400	0.012	0.016	
Z	0.380	0.720	0.015	0.028	
у	0.1		0.004		
θ	0°	8°	0°	8°	

- 1. Plastic or metal protrusions of 0.15mm maximum per side are not included.
- 2. This drawing is subject to change without notice.



SOP14 (2)

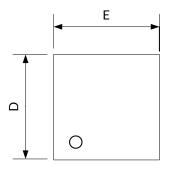


Completel.	Dimensions I	n Millimeters	Dimensions In Inches		
Symbol	Min	Max	Min	Max	
A (1)		1.750		0.069	
A ₁	0.100	0.250	0.004	0.010	
A ₂	1.250	1.450	0.049	0.057	
A ₃	0.:	25	0.0	10	
b _p	0.360	0.490	0.014	0.019	
С	0.190	0.250	0.007	0.010	
D ⁽¹⁾	8.550	8.750	0.340	0.350	
E (1)	3.800	4.000	0.150	0.160	
HE	5.800	6.200	0.228	0.244	
е	1.270		0.0	50	
L	1.05		0.0	41	
L _P	0.400	1.000	0.016	0.039	
Q	0.600	0.700	0.024	0.028	
Z	0.300	0.700	0.012	0.028	
У	0	.1	0.004		
θ	0°	8°	0°	8°	

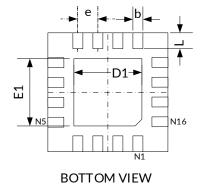
- Plastic or metal protrusions of 0.15mm maximum per side are not included.
 This drawing is subject to change without notice.

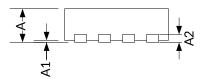


QFN3X3-16 (2)









SIDE VIEW

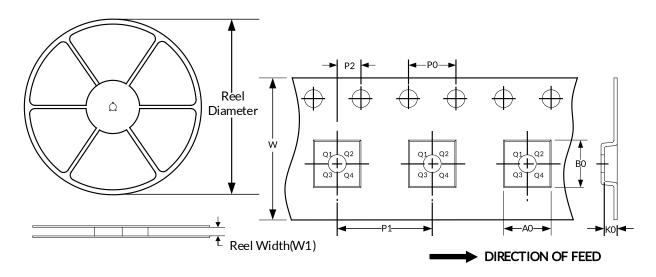
Compleal	Dimensions I	n Millimeters	Dimension	s In Inches
Symbol	Min	Max	Min	Max
A (1)	0.800	1.000	0.031	0.039
A1	0.000	0.050	0.000	0.002
A2	0.175	0.250	0.007	0.010
b	0.180	0.300	0.007	0.012
D (1)	2.950	3.050	0.116	0.120
D1	1.650	1.750	0.065	0.069
E ⁽¹⁾	2.950	3.050	0.116	0.120
E1	1.650	1.750	0.065	0.069
е	0.500	0.500 TYP) TYP
L	0.350	0.450	0.014	0.018

- Plastic or metal protrusions of 0.075mm maximum per side are not included.
 This drawing is subject to change without notice.



12 TAPE AND REEL INFORMATION REEL DIMENSIONS

TAPE DIMENSION



NOTE: The picture is only for reference. Please make the object as the standard.

KEY PARAMETER LIST OF TAPE AND REEL

Package Type	Reel	Reel	A0	B0	K0	P0	P1	P2	W	Pin1
	Diameter	Width(mm)	(mm)	Quadrant						
SOP14	13"	16.4	6.60	9.30	2.10	4.0	8.0	2.0	16.0	Q1
TSSOP14	13"	12.4	6.95	5.60	1.20	4.0	8.0	2.0	12.0	Q1
QFN3X3-16	13"	12.4	3.35	3.35	1.13	4.0	8.0	2.0	12.0	Q1

- 1. All dimensions are nominal.
- 2. Plastic or metal protrusions of 0.15mm maximum per side are not included.



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